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Adaptive Attitude Controller of Reentry Vehicles Based on Back-stepping Dynamic Inversion Method

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ABSTRACT: This paper presents an attitude control algorithm for a Reusable Launch Vehicle (RLV) Review History: with a low lift/drag ratio (L/D < 0.5), in presence of external disturbances, model uncertainties, Received: 16 May 2017 control output constraints and the thruster model. The main novelty of the proposed control strategy Revised: 22 September 2017 is a new combination of the attitude control methods including backstepping, dynamic inversion, Accepted: 24 October 2017 and adaptive control methods which will be called Backstepping-Dynamic inversion-Adaptive Available Online: 11 March 2018 (B.D.A) method. In the proposed method, a single control variable is considered as the bank angle while the angle of the attack and the side slip angle will be stabilized in their inherent value. The purpose of this control is the attitude control of the vehicle to track the commanded bank angle and keep the vehicle in the desired trajectory. Lyapunov stability analysis of the closed-loop system will be performed to guaranty the stability of the vehicle in the presence of constraints. Performance of the controller will be evaluated based on six Degrees of Freedom (6-DOF) model of the re-entry capsule. Also, the results of the proposed control algorithm will be compared with the Backstepping Dynamic inversion (B.D) control method

Keywords:

Attitude control Backstepping Dynamic inversion-Adaptive Reusable Launch Vehicle (RLV) Controller output constraint Thruster model

1-Introduction

Reusable Launch Vehicle (RLV) is widely used in Space missions. One of the most famous RLVs is the re-entry capsules with the low Lift/drag ratio (L/D) used in manned and payload return from space to the earth [1-3]. Designing a proper algorithm for the attitude control of an RLV in order to achieve various missions has generated many researches in this area [4-6]. Within re-entry phase, wide changes in aerodynamic characteristics, the significant coupling between the axes of the re-entry vehicle, external disturbances, and model uncertainties make it difficult to control the re-entry capsule. Also, the capsule must fly in a narrow flight corridor to limit mechanical load and heat flux that acted on the capsule [7-10]. Traditionally, different linear control methods have been proposed such as Linear Quadratic Regulator (LQR) and gain scheduling [11-13]. In these methods, several operating points are selected and the plant is linearized around these operating points. These methods are very simple, but they cannot guarantee global stability and, also, are timeconsuming in practice [14, 15]. In general, since the model of the re-entry vehicle is highly nonlinear, thus linear algorithms could not provide the stability in the re-entry phase. Thus, advanced control methods are very critical to meet safety, reliability and cost requirements for an RLV.

In order to improve the flight control performance of an RLV, many nonlinear control algorithms have been developed for attitude control of re-entry vehicles, such as dynamic inversion, sliding mode, adaptive, adaptive back-stepping, and adaptive sliding mode control methods. Nonlinear Dynamic Inversion (NDI) algorithm has been proposed to control a re-entry vehicle [16-18]. Implementation of the

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NDI method is very simple, and, also, in these papers a complete model of the re-entry capsule has been considered to design an attitude controller. However, this method is very sensitive to external disturbances and model uncertainties [19]. In order to solve this problem, versatile methods have been developed in recent years. In [20], the Sliding Mode Control (SMC) method has been proposed to control the attitude of a re-entry vehicle. Sliding mode control is one of the most robust control techniques and is insensitive to external disturbances and uncertainties [21]. However, the SMC method suffers from the chattering phenomenon [22, 23]. The boundary layer technique is used to alleviate the chattering [24]. Unfortunately, the robustness and accuracy of the SMC methods, within the boundary layer are no longer assured [9]. Thus, some advanced SMC methods have been developed to improve the performance of the SMC methods, such as nonsingular finite-time second-order sliding mode [9], and the second-order time-varying sliding mode [22]. In these references, the chattering problem of SMC has been solved, however, the actuator's dynamics and its output constraints have been neglected and not been considered in the design of the controller, which may reduce the controller performance in a practical situation. Geng Jie et al. [25] designed a finite-time sliding mode attitude controller for a re-entry vehicle and considered actuator's dynamics in their controller design, while the actuator's output constraints have not been considered; also, the translational motion of the reentry vehicle has not been considered in the controller design process. Thus, the equations of motions have been simplified. One of the other control method widely used in designing the attitude controller of re-entry vehicles is adaptive control, which does not have the chattering problem and also keeps a decent tracking performance in the presence of the

uncertainties [26]. Bai et al. [26] and Shi et al. [27] developed adaptive control methods to control the attitude of the reentry vehicle. Although their designed controller had a proper performance against disturbances and model uncertainties, they had not considered the actuator's model and the actuator output constraints. An adaptive controller has been designed to control the attitude of RLV by considering the uncertainties and the actuator's output constraints [28]; however, the actuator's model has not been considered. To solve these problems, an adaptive control has been proposed to consider the actuator's model and its constraints [29]; however, the reentry vehicle's dynamical model has been simplified, and the translational motion has been considered decoupled from the rotational motion. In some recent studies, the combination of control methods has been proposed to provide more effective attitude controller such as adaptive back-stepping and adaptive sliding mode. In the references [10, 30, 31], adaptive backstepping methods have been developed to control the attitude of RLV. They have considered the external disturbances, model uncertainties, and output constraints, however, all of them have not considered the actuator's model in their calculations. Wang L et al. [32] used combination of adaptive and sliding mode control methods, and also Wang Z et al. [33] designed an adaptive, backstepping, sliding mode controller. Both of these methods have had proper performances against external disturbances and model uncertainties, but they have not considered the actuator model in the controller's design process, and also they have used the simplified model of the re-entry motion in order to design the control algorithm. Generally, the main advantage of the mentioned methods is their suitable performance against external disturbance and uncertainties; but designing the controller by using these methods for a re-entry capsule is very complex. Thus, for designing the re-entry capsule's attitude controller needs many simplifying assumptions into the model. In addition, due to the actuator's model and output constraints, complexity of the controller design will be increased. Thus, in the most of these control methods, the actuator's dynamics is neglected, that may reduce the controller performance in a practical situation.

In this paper, a novel attitude control algorithm, named B.D.A will be designed for a reusable capsule with a low lift/drag ratio, (L/D ≈ 0.3). This algorithm included backstepping dynamic inversion and adaptive control methods which is used to control the attitude of the re-entry vehicle in the presence of external disturbances, uncertainties in the moment of inertia, the controller output constraints and the thruster model. In this method, the dynamic inversion algorithm will be designed to control the attitude angles by using the kinematic equations of the vehicle. Then, an adaptive mechanism will be designed to control the attitude angular rates by using dynamical equations. The main idea of the dynamic inversion method is to transform the dynamics of the nonlinear system into a linear one using the exact feedback of the system states or outputs. But, since dynamic inversion method is very sensitive to the model uncertainties, an adaptive algorithm will be designed by using the dynamical equations to control the attitude angular rates while it can keep a good tracking performance in the presence of the uncertainties [20]. A single control variable bank angle has been considered to control the attitude of the re-entry capsule while the angle of attack and side slip angle will be stabilized in their inherent

value in order to provide the required L/D. The simulation results will show that the proposed B.D.A method is effective to control the attitude of the re-entry capsule in the presence of the model uncertainties and external disturbances.

2- Re-entry capsule model

In this section, 6-DOF motion equations of the RLV during reentry phase will be described, and the control-oriented model will be derived to design the control algorithm. Dynamical equations of the motion of a re-entry capsule can be separated into the translational and rotational equations. Translational motion's equations can be written as follows [34]:

$$\dot{h} = V \sin\gamma , \ \dot{\mu} = \frac{V \cos\gamma \sin\chi}{\left(R_E + h\right)\cos\lambda} , \ \dot{\lambda} = \frac{V \cos\gamma \cos\chi}{\left(R_E + h\right)}$$
(1)

$$\dot{V} = -\frac{D\cos\beta}{m} + Y\sin\beta - g\sin\gamma + \Omega_E^2 \left(R_E + h\right)$$

 $\cos\lambda(\sin\gamma\cos\lambda - \cos\gamma\sin\lambda\cos\chi),$

$$\dot{\gamma} = -\frac{Dsin\sigma sin\beta - Ysin\sigma cos\beta + Lcos\sigma}{mV} -\left(\frac{g}{v} - \frac{V}{R_E + h}\right) cos\gamma + 2\Omega_E cos\lambda sin\chi$$

$$+\frac{\Omega_E^2 \left(R_E + h\right)}{V} cos\lambda \left(cos\gamma cos\lambda + sin\gamma sin\lambda cos\chi\right),$$

$$\dot{\chi} = \frac{Dcos\sigma sin\beta + Ycos\sigma cos\beta + Lsin\sigma}{mV cos\gamma} + \frac{V}{R_E + h}.$$
(2)

 $\cos\gamma\sin\chi\tan\lambda - 2\Omega_E(\tan\gamma\cos\lambda\cos\chi - \sin\lambda) +$

$$\frac{\Omega_E^2(R_E+h)}{V\cos\gamma}\sin\lambda\cos\lambda\sin\gamma.$$

Here, equation (1) describes the translational kinematic equations, and equation (2) describes the translational dynamic equations. $h, \mu, \lambda, V, \gamma, \chi$ are altitude, longitude, latitude, velocity, flight path angle, and heading angle respectively. L, D, Y are lift, drag, and side slip force and R_E, g, Ω_E are radius of the earth, gravity acceleration, and earth angular speed, respectively. Rotational motion's equations will be given as follows [35]:

$$\begin{split} \dot{\sigma} &= p \cos \alpha \cos \beta + q \sin \beta + r \sin \alpha \cos \beta - \\ \dot{\alpha} \sin \beta + \dot{\chi} \sin \gamma + \dot{\lambda} \sin \chi \cos \gamma \\ &- (\Omega_E + \dot{\mu}) (\cos \lambda \cos \chi \cos \gamma + \sin \lambda \sin \gamma), \\ \dot{\alpha} &= \frac{1}{\cos \beta} \begin{pmatrix} -p \cos \alpha \sin \beta + q \cos \beta - \\ r \sin \alpha \sin \beta - \sin \sigma (\dot{\chi} \cos \gamma - \dot{\lambda} \sin \chi \sin \gamma) \\ + (\Omega_E + \dot{\mu}) (\cos \lambda \cos \chi \sin \gamma - \sin \lambda \cos \gamma) \end{pmatrix}$$
(3)
$$- \cos \sigma (\dot{\gamma} - \dot{\lambda} \cos \chi - (\Omega_E + \dot{\mu}) \cos \lambda \sin \chi), \\ \dot{\beta} &= p \sin \alpha - r \cos \alpha - sin \begin{pmatrix} \dot{\gamma} - \dot{\lambda} \cos \chi - \\ (\Omega_E + \dot{\mu}) \cos \lambda \sin \chi \end{pmatrix} + \\ \cos \sigma \begin{pmatrix} \dot{\chi} \cos \gamma - \dot{\lambda} \sin \chi \sin \gamma \\ + (\Omega_E + \dot{\mu}) (\cos \lambda \cos \chi \sin \gamma + \sin \lambda \cos \gamma) \end{pmatrix}, \end{split}$$

$$\vec{\dot{\omega}} = -I^{-1} \vec{\omega} \times \vec{I} \vec{\omega} + I^{-1} \vec{M}.$$
(4)

where equations (3) and (4) describe the rotational kinematic and the rotational dynamic equations of re-entry capsule, respectively. The parameters α , β , σ , p, q, r are the angles of the attack, sideslip angle, bank angle, roll rate, pitch rate and yaw rate, respectively. Also I, M, ω represent the moment of inertia, capsule momentum, and angular speed matrix in the body frame, respectively, and are defined by the following relations

$$\vec{M} = \begin{bmatrix} m_{X} \\ m_{y} \\ m_{z} \end{bmatrix}, \quad I = \begin{bmatrix} I_{XX} & I_{XY} & I_{XZ} \\ I_{YX} & I_{YY} & I_{YZ} \\ I_{ZX} & I_{ZY} & I_{ZZ} \end{bmatrix},$$

$$\vec{\omega} = \begin{bmatrix} p \\ q \\ r \end{bmatrix}, \quad \vec{\omega} \times = \begin{bmatrix} 0 & -r & q \\ r & 0 & -p \\ -q & p & 0 \end{bmatrix}.$$
(5)

Equations (3) and (4) are the re-entry capsule motion's equations and will be used to design the controller algorithm.

3- Thruster Model

The general form of a linear thruster's model can be stated by equation (6) [36]:

$$T\dot{u} + u = v \tag{6}$$

where *u* is the vector of desired controller output, *v* is the vector of desired thruster input, and *T* is a diagonal square matrix with a positive time constant, hence $T=T^{T} > 0$ [36]. If *v* is equal to *u*, then the thruster is ideal and has no dynamics. However, thrusters have a dynamical model, which is usually presented by a transfer function, as follows:

$$G_{actuator} = \frac{u}{v} = \frac{1}{Ts+1} \tag{7}$$

Equation (7) is a conventional thruster's model which will be used for the control mission.

4-Backstepping Dynamic Inversion Adaptive Control (B.D.A)

In this section, the backstepping dynamic inversion adaptive (B.D.A) schemes will be presented. These schemes present a new combination of the attitude control methods, namely, backstepping dynamic inversion and adaptive control methods (B.D.A). The block diagram of B.D.A controller scheme has been shown in Figure 1. According to Figure 1, dynamic inversion principles will be used to transform the nonlinear equation (3) into a linear one in the first control loop. These linear equations will be controlled by designing a conventional PI controller.

In the first control loop, $\omega_c = [p_c \ q_c \ r_c]^T$ will be used to track the guidance command $\Omega_c = [\sigma_c \ \alpha_c \ \beta_c]^T$, and it is the output of the first control loop. Then, $\omega_c = [p_c \ q_c \ r_c]^T$ will be considered as the second control loop reference command. In the second control loop, the controller output u will be designed based on adaptive control method to track ω_c in the presence of external disturbances, model uncertainties, controller's output constraint and thruster's model. The guidance command vector $\Omega_c = [\sigma_c \ \alpha_c \ \beta_c]^T$ will be generated by using the guidance algorithm. In this paper, the Apollo capsule guidance algorithm has been used to generate the guidance commands.

4-1-B.D.A controller design

In this section, the proposed B.D.A control algorithm will be designed to control the re-entry capsule attitude in the presence of external disturbances, model uncertainties, controller's output constraint and thruster's model.

4-2-Dynamic inversion controller (first control loop)

Dynamic inversion controller will be designed to linearize equations (3). Thus, according to equations (3), we will have:

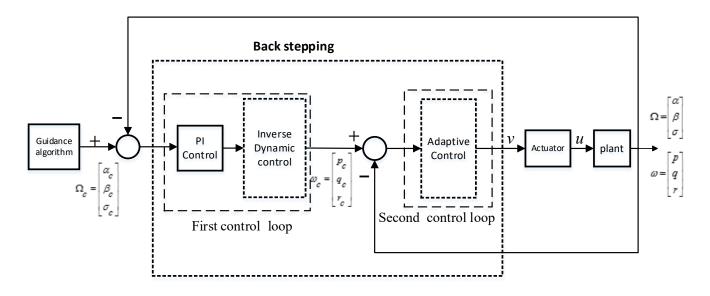


Figure1: Control scheme block diagram

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$$P_{C} = \frac{1}{\cos\alpha\cos\beta} \begin{pmatrix} \dot{\sigma} - q\sin\beta - r\sin\alpha\cos\beta \\ + \dot{\alpha}\sin\beta - \dot{\chi}\sin\gamma - \dot{\lambda}\sin\chi\cos\gamma \\ + (\Omega_{E} + \dot{\mu}) \\ (\cos\beta\cos\chi\cos\gamma + \sin\lambda\sin\gamma) \end{pmatrix} .$$

$$q_{C} = \frac{1}{\cos\beta} \begin{pmatrix} \cos\beta(\dot{\alpha} + \sin\sigma(\dot{\chi}\cos\gamma - \dot{\lambda}\sin\chi\sin\gamma \\ + (\Omega_{E} + \dot{\mu})(\cos\lambda\cos\chi\sin\gamma - \sin\lambda\cos\gamma) \end{pmatrix} + \\ \cos\sigma(\dot{\gamma} - \dot{\lambda}\cos\chi - (\Omega_{E} + \dot{\mu})\cos\lambda\sin\chi)). \tag{8}$$

 $pcos\alpha \sin\beta + rsin\alpha \sin\beta$)

$$r_{c} = -\frac{1}{\cos \alpha} (\dot{\beta} - p \sin \alpha + \sin(\dot{\gamma} - \dot{\lambda} \cos \chi - (\Omega_{E} + \dot{\mu}) \cos \lambda \sin \chi) - \cos \sigma (\dot{\chi} \cos \gamma - \dot{\lambda} \sin \chi \sin \gamma + (\Omega_{E} + \dot{\mu}) (\cos \lambda \cos \chi \sin \gamma + \sin \lambda \cos \gamma)))$$

where $\dot{\Omega} = \begin{bmatrix} \dot{\alpha} & \dot{\beta} & \dot{\sigma} \end{bmatrix}^T$ is the input vector of the first control loop and can be computed by using equation (9) as follows:

$$\dot{\Omega} = K_p(e) + K_I \int (e)dt, \tag{9}$$

where e is the attitude angle tracking error and will be considered as follows:

$$e = \left[\alpha_{c} - \alpha, \beta_{c} - \beta, \sigma_{c} - \sigma\right]^{T}.$$
(10)

 K_{P} and K_{I} are *PI* controller gains and will be defined, below:

$$K_{P} = \begin{bmatrix} K_{P1}, K_{P2}, K_{P3} \end{bmatrix}^{T},$$

$$K_{I} = \begin{bmatrix} K_{I1}, K_{I2}, K_{I3} \end{bmatrix}^{T}.$$
(11)

 $\omega_c = \begin{bmatrix} p_c & q_c & r_c \end{bmatrix}^T$ is the output vector of the first control loop and must be provided by the second control loop. In the next section, we will design the adaptive controller to control the attitude angular rate of the capsule.

4-3-Adaptive controller (second control loop)

In this section, an adaptive controller will be designed to control the attitude angular rate. Thus, equation (4) will be rewritten as follows:

$$I\dot{\omega} = -\omega \times I\omega + u. \tag{12}$$

By considering the external disturbances vector as $d = \begin{bmatrix} d_1 & d_2 & d_3 \end{bmatrix}^T$ and model uncertainties in the moment of the inertia matrix as ΔI and inserting them into equation (12), equation (12) can be written as follows:

$$(I + \Delta I)\dot{\omega} = -\omega \times (I + \Delta I)\omega + u + d, \tag{13}$$

$$\dot{\omega} = -I^{-1}\omega \times I\omega + I^{-1}u + D, \tag{14}$$

where:

$$D = I^{-1}(d - \Delta I\dot{\omega} - \omega \times \Delta I\omega).$$
⁽¹⁵⁾

and u is the controller output vector as follows:

$$u = \begin{bmatrix} M_x & M_y & M_z \end{bmatrix}, \tag{16}$$

where I and $\omega \times$ were defined in equation (5).

By considering the angular rate error as equation (17),

$$e = \omega - \omega_{com}, \tag{17}$$

we will have:

$$=\dot{\omega}-\dot{\omega}_{com}.$$
(18)

Now, the purpose is to design the controller output vector u by considering the controller output constraints. Based on the study in [10], we will consider the controller output constraints as $u = sat(u_0)$. By inserting it into equation (14), we will have:

$$\dot{\omega} = -I^{-1}\omega \times I\omega + I^{-1}sat(u_0) + D, \qquad (19)$$

where:

$$sat(u_0) = \delta_{u0} u_0. \tag{20}$$

 $sat(u_0) = [sat(u_{01}) \quad sat(u_{01}) \quad sat(u_{01})]^T$ and $sat(u_{0i})$ have been defined below:

$$sat(u_{0i}) = \begin{cases} u_{i\max} & u_{0i} \ge u_{i\max} \\ u_{0i} - u_{i\max} < u_{0i} < u_{i\max} \\ -u_{i\max} & u_{0i} < -u_{i\max} \end{cases}$$
(21)

According to equation (20), we have:

$$sat(u_{0i}) = \delta_{u0i}u_{0i}. \tag{22}$$

and can be written as follows:

$$\delta_{u0i} = \begin{cases} \frac{u_{i\max}}{u_{0i}} & u_{0i} \ge u_{i\max} \\ 1 & -u_{i\max} < u_{0i} < u_{i\max} \\ -\frac{u_{i\max}}{u_{0i}} & u_{0i} < -u_{i\max} \end{cases}$$
(23)

where $0 < \delta_{u0i} < 1$.

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(11

According to equations (19) and (20), we have:

$$\dot{\omega} = -I^{-1}\omega \times I\omega + I^{-1}\delta_{u0}u_0 + D.$$
⁽²⁴⁾

Inserting equation (24) into the equation (18), we have:

$$\dot{e} = -I^{-1}\omega \times I\omega + I^{-1}\delta_{u0}u_0 + D - \dot{\omega}_{com}.$$
(25)

Assume that there is a positive ρ satisfying $||D - \dot{\omega}_d|| \le \rho$. Since ρ must be estimated online, the adaptation law is used to update ρ . Based on the study in [10], the controller output can be computed as follows:

$$u_0 = -\left\|I\right\|^2 \lambda I^{-T} e^{-\frac{\hat{r}_{u0}(\left\|I^{-1}\omega \times I\omega\right\| + \hat{\rho} + \varepsilon)Ie}{\left\|e\right\|}}$$
(26)

and adaptation law for ρ and \hat{r}_{u0} will be computed as follows [10]:

$$\hat{\rho} = \mu_1 \left\| \boldsymbol{e} \right\| \tag{27}$$

$$\dot{\hat{r}}_{u_0} = \mu_2(\|I^{-1}\omega \times I\omega\| + \hat{\rho})\hat{r}_{u_0}^{-3}\|e\|$$
(28)

where $\lambda, \varepsilon, \mu_1, \mu_2, \hat{r}_{u0}(0), \hat{\rho}_{u0}(0) > 0$.

To prove the stability of the designed adaptive control law in equation (26), the Lyapunov function will be chosen as follows [10]:

$$V_1 = \frac{1}{2}e^T e + \frac{1}{2\mu_1}\tilde{\rho}^2 + \frac{1}{2\mu_2}\tilde{r}_{u0}^2, \qquad (29)$$

where $\tilde{\rho} = \hat{\rho} - \rho$ is the estimation error and $\tilde{r}_{u0} = \hat{r}_{u0}^{-1} - r_{u0}$. By using equations (27) and (28), the time derivative of V₁ is given by:

$$\begin{split} \dot{V}_{1} &= e^{T} \dot{e} + \frac{1}{\mu_{1}} \ddot{\rho} \dot{\dot{\rho}} + \frac{1}{\mu_{2}} \ddot{r}_{u0} \dot{r}_{u0}^{-2} \dot{\dot{r}}_{u0} \\ \dot{V}_{1} &= e^{T} \left(-I^{-1} \omega \times I \omega + I^{-1} \delta_{u0} u_{0} + D - \dot{\omega}_{com} \right) + \tilde{\rho} \| e \| \\ - \ddot{r}_{u0} \dot{r}_{u0} \left(\| I^{-1} \omega \times I \omega \| + \hat{\rho} \right) \| e \| \leq \| e \| \| I^{-1} \omega \times I \omega \| - \\ \ddot{r}_{u0} \| e \| \delta_{u0} \left(\| I^{-1} \omega \times I \omega \| + \hat{\rho} + \varepsilon \right) - \lambda \delta_{u0} \| e \|^{2} + \| e \| \rho \\ + \| e \| \tilde{\rho} - \ddot{r}_{u0} \dot{r}_{u0} \tilde{\rho} \| e \| \leq \| e \| \left(\| I^{-1} \omega \times I \omega \| + \rho \right) - \\ \lambda \delta_{u0} \| e \|^{2} - \| e \| \delta_{u0} \dot{r}_{u0} \left(\| I^{-1} \omega \times I \omega \| + \hat{\rho} \right) - \| e \| \delta_{u0} \dot{r}_{u0} \varepsilon \\ + \tilde{\rho} \| e \| - \| e \| \left(\| I^{-1} \omega \times I \omega \| + \hat{\rho} \right) + r_{u0} \dot{r}_{u0} \left(\| I^{-1} \omega \times I \omega \| + \hat{\rho} \right) \| e \| \\ &= -\lambda \delta_{u0} \| e \|^{2} - \delta_{u0} \dot{r}_{u0} \left(\| I^{-1} \omega \times I \omega \| + \hat{\rho} \right) \| e \| - \| e \| \delta_{u0} \dot{r}_{u0} \varepsilon \\ + r_{u0} \dot{r}_{u0} \left(\| I^{-1} \omega \times I \omega \| + \hat{\rho} \right) \| e \| \leq -\lambda \delta_{u0} \| e \|^{2} - \delta_{u0} \dot{r}_{u0} \| e \| \varepsilon \\ &\leq -\lambda_{\min} \left(\lambda \delta_{u0} \right) \| e \|^{2} - \| e \| \varepsilon. \end{split}$$

By integrating (27) with (28), we will have:

$$\hat{\rho}(t) = \hat{\rho}(0) + \mu_1 \int_0^t \left\| e \right\| dt,$$
(31)

$$\hat{r}_{u0}(t) = \hat{r}_{u0}(0) + \mu_2 \int_0^t (\left\| I^{-1} \omega \times I \omega \right\| + \hat{\rho}) \hat{r}_{u0}^3 \left\| e \right\| dt.$$
(32)

Thus, when $\|e\| \neq 0$, if the values of $\hat{\rho}(0), \hat{r}_{u0}(0), \mu_1, \mu_2$ are chosen appropriately, then there is a time, say t_2 , to make $\hat{\rho}(t) \ge \rho(t)$ and $\hat{r}_{u0}(t) \ge \hat{r}_{u0}^{-1}(t)$. Also, $\tilde{\rho}(t) \ge 0$ and $\tilde{r}_{u0} \le 0$. According to equation (30), we have:

$$e^{T}\dot{e} + \tilde{\rho} \|e\| - \tilde{r}_{u0}\hat{r}_{u0} (\|I^{-1}\omega \times I\omega\| + \hat{\rho})\|e\|$$

$$\leq -\varepsilon \|e\|.$$
(33)

Therefore, it is clear that $\dot{V_1} \le 0$ and *e* converges to the origin in the finite time.

4- 4- Stability Proof of the B.D.A Algorithm in presence of the thruster model

In this section, the stability proof of the proposed B.D.A controller will be achieved in presence of the thruster's model which was mentioned in equation (6). Since the thruster has dynamics, it must be considered in the design of control algorithm and the stability proof of B.D.A must be driven again. Therefore, equation (6) will be rewritten as follows:

$$T\dot{u} = v - u,\tag{34}$$

and the thrust error will be considered as follows [36]:

$$z = u - \tau. \tag{35}$$

which z is the error between desired controller output u and the real torque τ . Equation (35) can be rewritten as follows:

$$\dot{z} = \dot{u} - \dot{\tau},\tag{36}$$

$$T\dot{z} = T\dot{u} - T\dot{\tau}.$$
(37)

According to equations (34) and (37), we have:

$$T\dot{z} = v - u - T\dot{\tau}.\tag{38}$$

Then, a new Lyapunov function has been chosen as follows:

$$V_2 = V_1 + \frac{1}{2}z^T T z. ag{39}$$

Thus, the time derivative of equation (39) can be computed as follows:

$$\dot{V}_2 = \dot{V}_1 + z^T T \dot{z}.$$
 (40)

By considering equations (38) and (40), we will have:

$$\dot{V}_2 = \dot{V}_1 + z^T (v - u - T\dot{\tau}).$$
 (41)

Then, a desired thruster input *v* will be designed as follows:

$$v = -Kz + T\dot{\tau} + u \tag{42}$$

where $K = K^T > 0$ is the feedback gain matrix.

According to equation (41), equation (42) can be written as follows:

$$\dot{V}_{2} = \dot{V}_{1} + z^{T} (-Kz + T \dot{\tau} + u - u - T \dot{\tau}) = \dot{V}_{1} - z^{T} K z \le 0.$$
(43)

From equation (43), it is clear that $\dot{V_2} < 0$ for $z \neq 0$. Investigation of V_2 shows that $V_2 > 0$ for $z \neq 0$, and also that $V_2 \rightarrow \infty$ when $z \rightarrow \infty$. Besides, this is an autonomous system, owing to the fact that the system does not depend on time. These three criteria state that the controller is globally asymptotically stable according to Lyapunov's direct method [37]. Hence, the adaptive controller is globally stable in the presence of the thruster model.

Remark 1: According to (26), the controller output will cause chattering. By defining a boundary layer around e, the chattering phenomenon can be omitted [10]. Thus, the controller output in equation (26) can be computed as follows:

$$u_{0} = \begin{cases} -\|I\|^{2} \lambda I^{-T} e - \frac{\hat{r}_{u0}(\|I^{-1} \omega \times I \omega\| + \hat{\rho} + \varepsilon)Ie}{\|e\|} & \text{if } \|e\| > \varepsilon_{u0} \\ -\|I\|^{2} \lambda I^{-T} e - \frac{\hat{r}_{u0}(\|I^{-1} \omega \times I \omega\| + \hat{\rho} + \varepsilon)Ie}{\varepsilon_{u0}} & \text{if } \|e\| \le \varepsilon_{u0} \end{cases}$$
(44)

Remark 2: In real conditions, ||e|| cannot be exactly zero. Thus, based on equations (27) and (28), $\hat{\rho}$ and \hat{r}_{u0} may be increased boundlessly. In order to solve this problem, the adaptation laws have been modified as follows [10].

$$\dot{\hat{\rho}} = \begin{cases} \mu_1 \|\boldsymbol{e}\| & \text{if } \|\boldsymbol{e}\| > \varepsilon_{u_0} \\ 0 & \text{if } \|\boldsymbol{e}\| \le \varepsilon_{u_0} \end{cases}$$

$$\left\{ \mu_1 \left(\|\boldsymbol{I}^{-1} \boldsymbol{\omega} \times \boldsymbol{I} \boldsymbol{\omega}\| + \hat{\boldsymbol{\omega}} \right) \hat{\boldsymbol{\varepsilon}}^{-3} \|\boldsymbol{e}\| & \text{if } \|\boldsymbol{e}\| > \varepsilon_{u_0} \end{cases}$$

$$(45)$$

$$\hat{r}_{u_0} = \begin{cases} \mu_2(\|I \ \omega \times I\omega\| + \rho)\hat{r}_{u_0} \ \|e\| \ \|e\| > \varepsilon_{u_0} \\ 0 \ & if \ \|e\| \le \varepsilon_{u_0} \end{cases}$$
(46)

In equations (9) and (44), the larger values of K_p , λ , ε present a faster convergence of the system; however, the controller output will be increased. Also, larger value of μ_1 , μ_2 will cause faster convergence of $\hat{\rho}$, \hat{r}_{u0} ; however, the controller output will be increased, as well. Thus, there must be a tradeoff between accuracy, faster convergence of the system, and controller output.

4-5-Dynamic inversion controller (second control loop) design for B.D control algorithm.

In this section, a dynamic inversion controller will be designed for the second control loop to track reference commands that are generated in the first loop. If both two control loop are designed based on dynamic inversion, the control algorithm will be titled the B.D. The B.D.A controller and the B.D controller have been designed independently from each other. The B.D control algorithm is designed based on reference [16]. Similar to 4-2, to design dynamic inversion controller for the second control loop, equation (12) can be written as follows [16]:

$$\vec{u} = I\vec{\omega} + \vec{\omega} \times I\vec{\omega} \tag{47}$$

where $\dot{\omega} = \begin{bmatrix} \dot{p} & \dot{q} & \dot{r} \end{bmatrix}^T$ is the input vector of the second control loop and can be computed by employing equation (48) as follows:

$$\dot{\omega} = K_{p2}(e_2) + K_{I2} \int (e_2) dt, \tag{48}$$

where e_2 is the attitude angle rate error vector and can be computed as follows:

$$e_2 = [p_C - p, q_C - q, r_C - r]^T.$$
(49)

 K_{PP} , and K_{PP} are the PI controller gains defined as follows:

$$K_{P2} = \begin{bmatrix} K_{P21}, K_{P22}, K_{P23} \end{bmatrix}^{T},$$

$$K_{I2} = \begin{bmatrix} K_{I21}, K_{I22}, K_{I23} \end{bmatrix}^{T}.$$
(50)

Therefore, equation (8) can be used to design the first control loop in both B.D and B.D.A schemes. Equation (47) is used to design the second control loop in the B.D scheme and equations (44), (45) and (46) are utilized to design the second control loop in the B.D.A scheme. The simulation results will be presented in the next section.

5- Simulation And Results Analysis

In this section, simulation of the designed control algorithm (B.D.A) will be performed based on a 6-DOF dynamics of the capsule and, next, the results analysis will be presented. Finally, the comparison between the B.D.A controller and the B.D controller will be carried out to evaluate the proposed B.D.A controller.

Simulation parameters including the inertia matrix (I), external disturbances (d), inertia matrix uncertainties (ΔI) , controller output constraints (u_{\max}, u_{\min}) and the time constant (T) of the thruster model have been summarized in the Table 1.

The initial flight conditions of the capsule have been presented in the Table 2

Ta	ble	1:	Simu	lation	para	ameters	\$.
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Param- eter	Definition	Value		
Ι	Moment of inertia matrix of the capsule	$\begin{bmatrix} 7985.5 & 59.7 & -503 \\ 59.7 & 7150.6 & -32.5 \\ -503 & -32.5 & 6456.4 \end{bmatrix} kg - m^2$		
d	External disturbance matrix	$\begin{bmatrix} 1 + \sin(\frac{\pi t}{125}) + \sin(\frac{\pi t}{250}) \\ 1 + \sin(\frac{\pi t}{125}) + \sin(\frac{\pi t}{250}) \\ 1 + \sin(\frac{\pi t}{125}) + \sin(\frac{\pi t}{250}) \end{bmatrix} \times 10^2 \text{ N.m}$		
ΙΔ	Uncertainties in the Moment of inertia matrix	0.2 I		
u _{max}	Maximum controller output	5000 N.m		
$u_{_{min}}$	Minimum controller output	-5000 N.m		
Т	Thruster model time constant	0.05s		

Table 2: Initial conditions.

Param- eter	Definition	Value
h_0	Altitude	121920 m
v_0	Velocity	7832 m/s
μ_0	Longitude	49 deg
λ_{0}	Latitude	35 deg
${\gamma}_0$	Flight path angle	-4.3 deg
${\mathcal X}_0$	Heading angle	136 deg
$lpha_{_0}$	Angle of attack	-15 deg
eta_0	Side slip angle	0 deg
$\sigma_{_0}$	Bank angle	0 deg
p_0	Roll rate	0(deg/s)
${q_0}$	Pitch rate	0(deg/s)
r ₀	Yaw rate	0(deg/s)

Parameter	μ_1, μ_2	Е	λ	\mathcal{E}_{u0}	$\boldsymbol{K}_{p1} = \boldsymbol{K}_{p2} = \boldsymbol{K}_{p3}$	$K_{I1} = K_{I2} = K_{I3}$	$\hat{r}_{u0}(0)$	$\hat{ ho}(0)$
Value	0.1	4	0.2 . diag (1,0.5,0.5)	0.05	0.4	0.01	0	0

Table 3: Parameters of B.D.A controller

5- 1- Simulation and control performance analysis of B.D.A controller

In this section, the simulation results and performance analysis of the B.D.A controller will be presented. The parameters of the B.D.A controller are summarized in Table 3.

The performance of the controller to track the guidance commands is shown in Figure 2. In this case, the tracking quality seems satisfactory and is in a great compliance with the desired values.

Figure 3 shows the guidance commands tracking errors. It shows that the maximum tracking error of the attack angle and side slip angle are less than 1 degree. The maximum tracking error of the bank angle is higher than 100 degrees. This is due to the inertia of the capsule when the guidance

commands change. The tracking errors can be reduced by augmenting the controller outputs constraints levels.

In Figure 4, the rates of angular velocity have been illustrated. It is clear that reference commands have been tracked properly by applying the proposed control algorithm.

Figure 5 presents the controller outputs. It is obvious that the value of the controller's outputs is less than the controller's output constraints levels. Because of the stabilization of the re-entry capsule in pitch and yaw axis, a small controller output has been generated.

The load factor, landing accuracy, and L/D ratio are presented in Figure 6 that illustrates that the value of the load is less than 5g and landing accuracy is less than 500 m. Also, the L/D ratio is kept around 0.3.

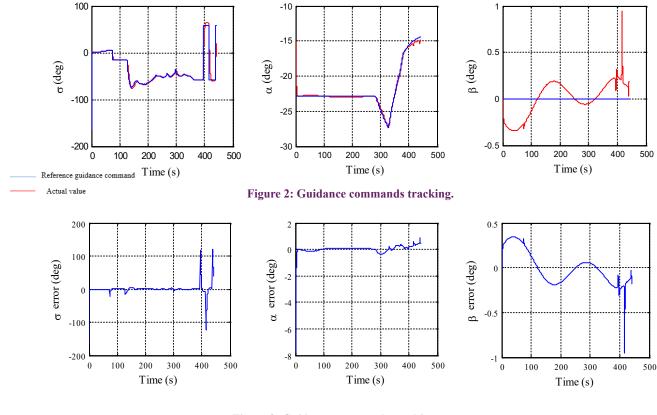
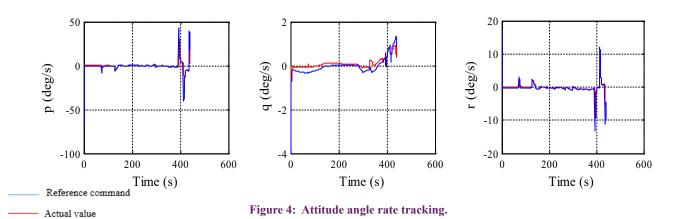
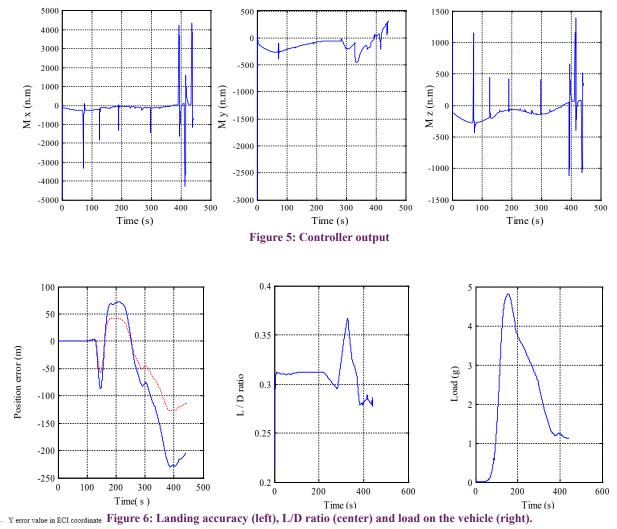


Figure 3: Guidance commands tracking error.



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X error value in ECI coordinate

In order to evaluate the performance of the B.D.A controller, the back-stepping dynamic inversion (B.D) controller has been designed to be used in the second control loop and the results will be compared with the proposed B.D.A controller.

5-2-Simulation and control performance analysis of B.D controller

In this section, simulation of the B.D control algorithm will be performed based on a 6-DOF dynamical model of the capsule and the results analysis will be presented. The initial conditions are the same as conditions in the B.D.A controller which have been presented in the Table 2. The B.D controller parameters have been summarized in Table 4.

Figure 7 shows that the B.D controller performance is satisfactory as it tracks the guidance commands.

Figure 8 depicts reference angle tracking error by applying the B.D algorithm. The angle errors change according to variations of the reference guidance commands. These errors are less than 1 degree except in the cases where the reference guidance commands change abruptly due to the high inertia of the re-entry vehicle.

Reference angular rates tracking has been presented in Figure 9 which illustrates fine angular rate reference command tracking. Because of the coupling between re-entry capsule

axes, there is a close relation between roll rate and yaw rate of the re-entry capsule which causes yaw rate to change when roll rate varies.

The B.D controller outputs are shown in Figure 10. It shows that the controller outputs have exceeded controller output constraint (5000 N.m) in roll and yaw axes. In roll axis, the controller output is about -40000 N.m which is higher than the controller output constraint.

Table 4: Parameters of the B.D controller

Parameter	Value	Parameter	Value
\boldsymbol{K}_{p1}	0.4	K ₁₃	0.04
K_{p2}	0.5	K _{p21}	5
K_{p^3}	0.3	K _{p22}	7
K ₁₁	0.05	K _{p23}	4
K ₁₂	0.05	$K_{I21} = K_{I22} = K_{I23}$	0

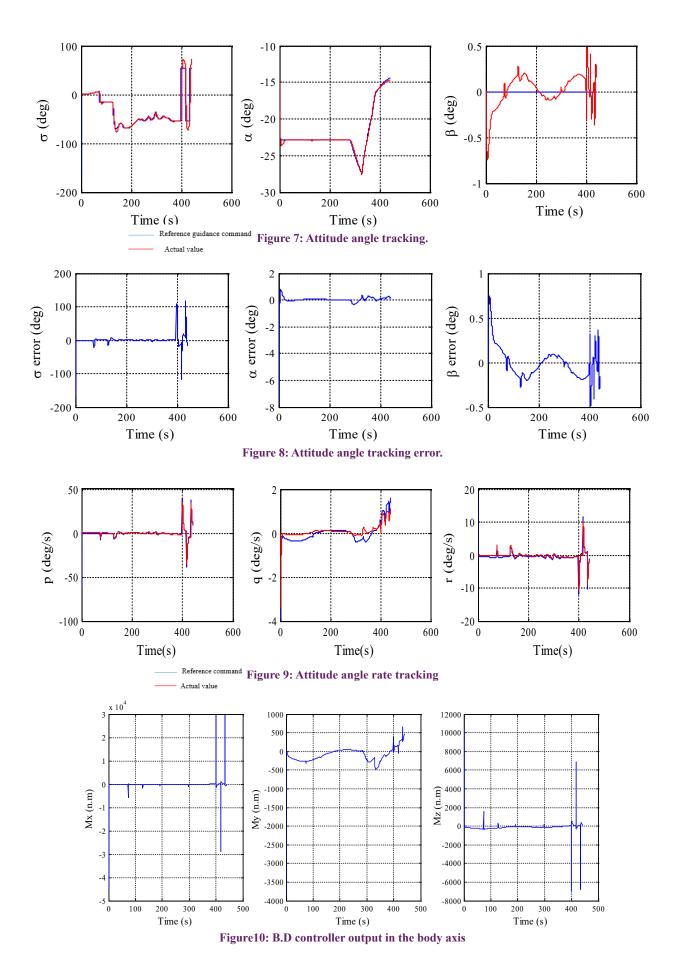


	Table 5: Comparison between the B.D and B.D.A controllers.						
Controller method	Control method of first control loop	Control method of second con- trol loop	Accuracy of angle of attack and side slip angle (deg)	Mx (N.m)	My (N.m)	Mz (N.m)	
B.D.A	Dynamic inversion	Adaptive	<10	$-5000 < M_{\chi} < 5000$	$-3000 < M_{\gamma} < 500$	$-500 < M_{Z} < 1500$	
B.D	Dynamic inversion	Dynamic inversion	<10	$-50000 < M_{\chi} < 30000$	$-4000 < M_{\gamma} < 1000$	$-8000 < M_z < 12000$	

According to the simulation results of the B.D.A and B.D controller, a comparison between B.D.A and B.D controllers can be summarized as follows:

1. Both B.D.A and B.D controllers provide a stable control system in the presence of the external disturbances and uncertainties in the moment of inertia matrix. The angle of attack and side slip angle errors are less than 1 degree.

2. The controller's output in the B.D controller is greater than 40000 N.m in the roll axis while the B.D.A controller's output is less than 5000 N.m in the same conditions.

3. The controller's output in the B.D controller is greater than 8000 N.m in the yaw axis while the B.D.A controller's output is less than 5000 N.m in the same conditions.

4. The B.D controller is more sensitive to the model uncertainties than the B.D.A controller. Hence, the B.D.A controller is more appropriate than the B.D controller in the phase of re-entry.

The results of comparison between the B.D and B.D.A controllers are summarized in Table 5.

According to Table 5, it is clear that the B.D.A controller is more effective than the B.D controller and generates lower controller's output in the same conditions.

6- Conclusion

In this paper, a new attitude control algorithm, called B.D.A, has been designed based on backstepping dynamic inversion and adaptive control methods to control the attitude of a low L/D re-entry vehicle in the presence of the external disturbances, uncertainties in the moment of inertia, controller's output constraints and the thruster's model. In this method, the dynamic inversion algorithm has been designed to control the attitude angles using the kinematic equations of the vehicle. Then, an adaptive mechanism has been designed to control the attitude angular rates using the dynamical equations.

A single control variable bank angle has been considered to control the attitude of the re-entry capsule. The simulation results have shown that the B.D.A controller can provide a stable guidance command tracking. The proposed controller's outputs are less than 5000 N.m. The B.D.A controller results have been compared with those of the B.D controller to evaluate the performance of the B.D.A controller. The simulation results have shown that the B.D controller is very sensitive to the model uncertainties and the external disturbances and it generates too high (higher than 40000 N.m) controller's output while the B.D.A controller output is less than 5000 N.m in the same conditions. It means that, in comparison with the B.D controller, the proposed B.D.A method is more effective to control the attitude of the re-entry capsule in the presence of the model uncertainties and the external disturbances.

Nomenclature

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Symbols	Definitions	Symbols	Definitions
h	Altitude	L	Lift force
V	Velocity	D	Drag force
μ	Longitude	Y	Side slip force
λ	Latitude	т	Capsule mass
γ	Flight path angle	Ι	Moment of inertia matrix
χ	Heading angle	u	Controller output
α	Angle of attack	R_{E}	Radios of earth
β	Side slip angle	$\Omega_{_E}$	Angular velocity of earth
σ	Bank angle	Т	Thruster mode time constant
р	Roll rate	е	Tracking error
q	Pitch rate	d	External disturbance
r	Yaw rate	ΔI	Uncertainty in moment of inertia matrix
Ω	Vector of at- titude angle	V	Desired thruster input
ω	Vector of at- titude angle rate	τ	Real torque of thruster

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